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$$\frac{H_1}{\left| \underline{\mathbf{Y}}(u) - \underline{\mathbf{H}}^{-1}\underline{\mathbf{s}}_1 \right|^2} \mathop{<}_{\substack{> \\ H_0}} \left| \underline{\mathbf{Y}}(u) - \underline{\mathbf{H}}^{-1}\underline{\mathbf{s}}_0 \right|^2$$

$$\Leftrightarrow |\underline{\mathbf{Y}}(u)|^{2} + |\underline{\mathbf{H}}^{-1}\underline{\mathbf{s}}_{1}|^{2} - 2\underline{\mathbf{Y}}(u)^{T}\underline{\mathbf{H}}^{-1}\underline{\mathbf{s}}_{1} \lesssim |\underline{\mathbf{Y}}(u)|^{2} + |\underline{\mathbf{H}}^{-1}\underline{\mathbf{s}}_{0}|^{2} - 2\underline{\mathbf{Y}}(u)^{T}\underline{\mathbf{H}}^{-1}\underline{\mathbf{s}}_{0}$$

$$\Leftrightarrow \ \, \underline{Y}(u)^{T}[\underline{H}^{-1}\underline{s}_{1} - \underline{H}^{-1}\underline{s}_{0}] \overset{H_{1}}{\underset{H_{0}}{\swarrow}} T = \frac{|\underline{H}^{-1}\underline{s}_{1}|^{2} - |\underline{H}^{-1}\underline{s}_{0}|^{2}}{2}$$

In principle we can stop here. In addition this can be simplified by

substitution as follows

$$\mathbf{x}^T \mathbf{\underline{H}}^{-1} \mathbf{\underline{H}}^{-1} (\underline{\mathbf{s}}_1 - \underline{\mathbf{s}}_0) \stackrel{>}{<} T$$

since $\underline{\mathbf{H}}\underline{\mathbf{H}}^T = \underline{\mathbf{K}}_N$

$$\underline{\mathbf{x}}^{T}\underline{\mathbf{K}}_{N}^{-1}(\underline{\mathbf{s}}_{1}-\underline{\mathbf{s}}_{0}) \stackrel{H_{1}}{<} T$$

$$H_{0}$$

where

$$T = \frac{|\underline{\mathbf{H}}^{-1}\underline{\mathbf{s}}_{1}|^{2} - |\underline{\mathbf{H}}^{-1}\underline{\mathbf{s}}_{0}|^{2}}{2}$$
$$= \frac{\underline{\mathbf{s}}_{1}^{T}\underline{\mathbf{K}}_{N}^{-1}\underline{\mathbf{s}}_{1} - \underline{\mathbf{s}}_{0}^{T}\underline{\mathbf{K}}_{N}^{-1}\underline{\mathbf{s}}_{0}}{2}$$

i.e. the new version of the decision mechanism looks like this.

Example:

$$H_1$$
: $\underline{X}(u) = \begin{bmatrix} 0 \\ 10 \end{bmatrix} + N(u)$

$$H_0$$
: $\underline{\mathbf{X}}(u) = \begin{bmatrix} 0 \\ -10 \end{bmatrix} + N(u)$

where

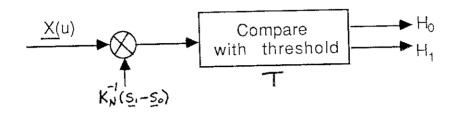
$$\underline{\mathbf{m}}_N = \underline{\mathbf{0}} \ , \ \underline{\mathbf{K}}_N = \left[\begin{array}{cc} \mathbf{4} & \mathbf{2} \\ \mathbf{2} & \mathbf{7} \end{array} \right]$$

The above example has the following pictorial representation. However $\underline{K}_N \neq \underline{I}$

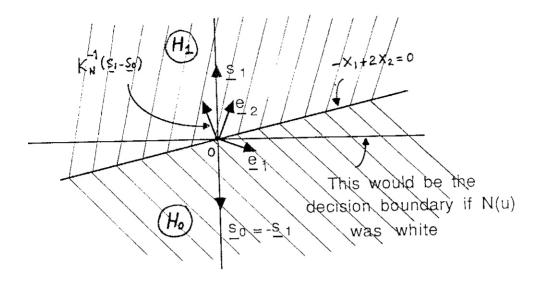
$$\Rightarrow \underline{\mathbf{K}}_{N}^{-1} = \frac{1}{24} \left[\begin{array}{cc} 7 & -2 \\ -2 & 4 \end{array} \right]$$

and form

$$\underline{\mathbf{K}}_{N}^{-1}(\underline{\mathbf{s}}_{1} - \underline{\mathbf{s}}_{0}) = \frac{1}{24} \begin{bmatrix} 7 & -2 \\ -2 & 4 \end{bmatrix} \begin{bmatrix} 0 \\ 20 \end{bmatrix} = \begin{pmatrix} \frac{5}{3} \end{pmatrix} \begin{bmatrix} -1 \\ 2 \end{bmatrix}$$



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Also,
$$\underline{\mathbf{s}}_1^T \underline{\mathbf{K}}_N^{-1} \underline{\mathbf{s}}_1 = \underline{\mathbf{s}}_0^T \underline{\mathbf{K}}_N^{-1} \underline{\mathbf{s}}_0$$
 $\Rightarrow T = 0$

So the decision rule is

$$\underline{\mathbf{X}}^{T}(u) \begin{pmatrix} \frac{5}{3} \end{pmatrix} \begin{bmatrix} -1 \\ 2 \end{bmatrix} \stackrel{H_{1}}{\stackrel{>}{\stackrel{>}{\stackrel{}{\sim}}}} 0$$

We can eliminate the constant factor to get the following decision rule

$$\Rightarrow -X_1(u) + 2X_2(u) \begin{vmatrix} H_1 \\ > \\ < 0 \\ H_0 \end{vmatrix}$$

To understand the *slope* of the decision boundary we write the K-L expansion of the noise:

$$egin{aligned} & \mathbb{K}_N = \left[egin{array}{cc} 4 & 2 \\ 2 & 7 \end{array}
ight] \ & \lambda_1 = 3 \Rightarrow \underline{e}_1 = rac{1}{\sqrt{5}} \left[egin{array}{cc} 2 \\ -1 \end{array}
ight] \ & \lambda_2 = 3 \Rightarrow \underline{e}_2 = rac{1}{\sqrt{5}} \left[egin{array}{cc} 1 \\ -2 \end{array}
ight] \end{aligned}$$

i.e. $\underline{N}(u) = \sqrt{3}W_1(u)\underline{e}_1 + \sqrt{8}W_2(u)\underline{e}_2$. In order to perform the above, we have assumed that \underline{K}_N^{-1} exists, that is the components of the noise are *not* linearly dependent. What would happen however if \underline{K}_N were singular, i.e. if $\lambda_i = 0$?

This is the case of singular detection, in which a perfect decision can be made!

Example: Let us use the same signals as before

$$\begin{bmatrix} 0 \\ 10 \end{bmatrix}, \begin{bmatrix} 0 \\ -10 \end{bmatrix}$$

$$\underline{K}_{N} = \begin{bmatrix} 1 & 3 \\ 3 & 9 \end{bmatrix} \Rightarrow \begin{cases} \lambda_{1} = 0 \\ \lambda_{2} = 0 \end{cases}$$

$$\Rightarrow \underline{e}_{1} = \frac{1}{\sqrt{10}} \begin{bmatrix} -3 \\ 1 \end{bmatrix} ; \underline{e}_{2} = \frac{1}{\sqrt{10}} \begin{bmatrix} 1 \\ 3 \end{bmatrix}$$

The two noise components are linearly dependent, i.e.

$$\underline{\mathbf{N}}(u) = \sqrt{\lambda_2} W_2(u) \underline{\mathbf{e}}_2 = W_2(u) \begin{bmatrix} 1 \\ 3 \end{bmatrix} = \begin{bmatrix} W_2(u) \\ 3W_2(u) \end{bmatrix}$$

So if we project along e_1 , which is the eigenvector with zero eigenvalue,

we can make an unambiguous decision (i.e. with no errors) Since

$$\Rightarrow \underline{e}_1 = \frac{1}{\sqrt{10}} \left[\begin{array}{c} -3\\1 \end{array} \right]$$

the decision rule is

$$(\underline{e}_1, \underline{X}) = -3X_1 + X_2 \stackrel{>}{<} 0$$

We have the following general statement:

Suppose \underline{K}_N is singular, with $\lambda_1 = 0$ and some eigenvactor \underline{e}_1 corresponding to λ_1 . Then

$$\underline{N}(u) = \sum_{j=2}^{n} W_{j}(u) \sqrt{\lambda_{j}} \underline{e}_{j}$$

$$\underline{X}(u) = \underline{s}_{i} + \sum_{j=2}^{n} W_{j}(u) \sqrt{\lambda_{j}} \underline{e}_{j}$$

$$\Rightarrow \underline{X}^{T}(u) \underline{e}_{1} = \underline{s}_{i}^{T} \underline{e}_{1} + \sum_{j=2}^{n} \sqrt{\lambda_{j}} W_{j}(u) \underline{e}_{j}^{T} \underline{e}_{1}$$

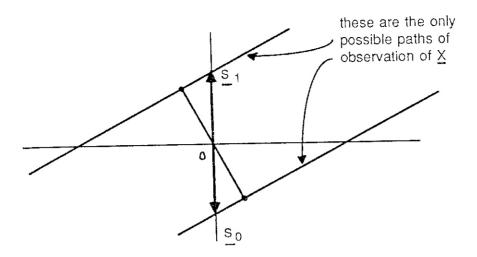
but $\underline{\mathbf{e}}_{j}^{T}\underline{\mathbf{e}}_{1}$ is zero by the orthogonality of the $\underline{\mathbf{e}}_{i}$'s

$$\Rightarrow \underline{\mathbf{X}}^T(u)\underline{\mathbf{e}}_1 = \underline{\mathbf{s}}_i^T\underline{\mathbf{e}}_1$$

that means noiseless reception! Consequently if

$$\underline{\mathbf{s}}_{1}^{T}\underline{\mathbf{e}}_{1} \neq \underline{\mathbf{s}}_{0}^{T}\underline{\mathbf{e}}_{1}$$

a perfect decision is possible.



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